

**Texas A&M University
Department of Electrical and Computer Engineering**

ECEN 474/704 – (Analog) VLSI Circuit Design

Fall 2016

Exam #3

Instructor: Sam Palermo

- Please write your name in the space provided below
- Please verify that there are 5 pages (1 blank) in your exam
- You may use one double-sided page of notes and equations for the exam
- Good Luck!

Problem	Score	Max Score
1		35
2		35
3		30
Total		100

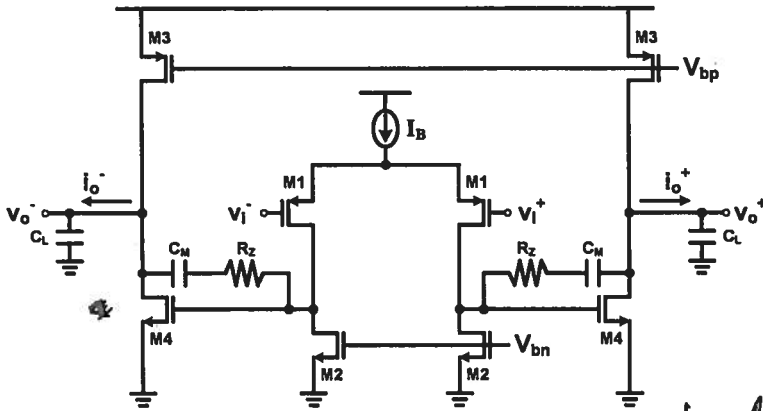
Name: SAM PALERMO

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Problem 1 (35 points)

For the fully-differential amplifier below, assume all transistors are operating in saturation and obtain expressions for the following:

- a) Small-signal differential transconductance, $(i_o^+ - i_o^-)/(v_i^+ - v_i^-)$.
- b) Fully differential amplifier DC gain, $A_{vd} = (v_o^+ - v_o^-)/(v_i^+ - v_i^-)$.
- c) The amplifier's two main poles. Note, it's OK to neglect the transistor capacitors here.
- d) What is the value of R_z that sets the main compensation zero to infinity?



a. $G_m = \left(\frac{g_{m1}}{g_{o1} + g_{o2}} \right) g_{m4}$

$R_{out} = \frac{1}{g_{o3} + g_{o4}}$

b. $A_v = G_m R_{out}$

$= \frac{g_{m1} g_{m4}}{(g_{o1} + g_{o2})(g_{o3} + g_{o4})}$

c. $\omega_{p1} \approx - \frac{(g_{o3} + g_{o4})(g_{o1} + g_{o2})}{C_m g_{m4}}$

$\omega_{p2} \approx - \frac{g_{m4}}{C_L}$

d. $\omega_z = \frac{1}{\left(\frac{1}{g_{m4}} - R_z \right) C_m} = \infty$

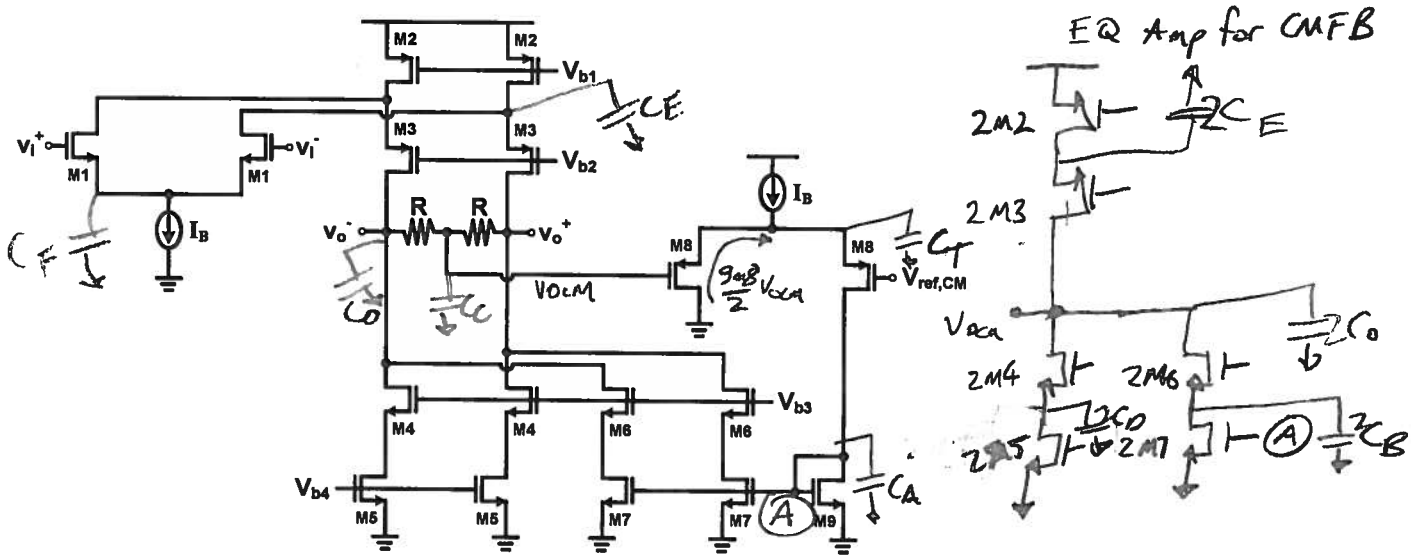
$\Rightarrow \frac{1}{g_{m4}} - R_z = 0$

$R_z = \frac{1}{g_{m4}}$

Problem 2 (35 points)

For the fully differential amplifier with common-mode feedback (CMFB) below, assume all transistors are operating in saturation, and obtain the following:

- Neglecting the CMFB network, give an expression for the fully differential amplifier DC gain, $A_{vd} = (v_o^+ - v_o^-) / (v_i^+ - v_i^-)$.
- Give an expression for the CMFB loop DC gain.
- Give expressions for the poles of the CMFB loop. Note, it's OK here to state this as a function of an effective capacitance at a certain node, but make sure to appropriately label the nodes.



a.
$$A_{vd} = g_{m1} \left[g_{m3} r_{o3} (r_{o2} || r_{o4}) || g_{m4} r_{o4} r_{o5} || g_{m6} r_{o6} r_{o7} \right]$$

b. Gain from v_{ocm} to $v_A = \frac{g_{m8}}{2} \left(\frac{1}{g_{m9}} \right) = \frac{g_{m8}}{2g_{m9}}$

CMFB Loop Gain =
$$\left(\frac{g_{m8}}{2g_{m9}} \right) (-2g_{m7}) \left[2g_{m3} \left(\frac{r_{o5}}{2} \right) \left(\frac{r_{o2}}{2} \right) || 2g_{m4} \left(\frac{r_{o4}}{2} \right) \left(\frac{r_{o5}}{2} \right) || 2g_{m6} \left(\frac{r_{o6}}{2} \right) \left(\frac{r_{o7}}{2} \right) \right]$$

CMFB Loop Gain =
$$-\frac{g_{m8} g_{m7}}{g_{m9}} \left[\frac{g_{m3} r_{o3} r_{o2}}{2} || \frac{g_{m4} r_{o4} r_{o5}}{2} || \frac{g_{m6} r_{o6} r_{o7}}{2} \right]$$

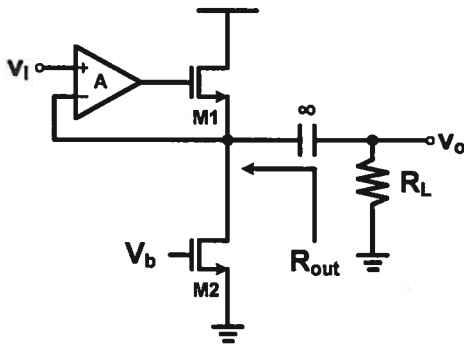
c. Poles at
$$\left(g_{m3} r_{o3} r_{o2} || g_{m4} r_{o4} r_{o5} || g_{m6} r_{o6} r_{o7} \right) C_a, \frac{2g_{m8}}{C_T}, \frac{g_{m9}}{C_A}$$

$$\frac{g_{m6}}{C_B}, \frac{g_{m4}}{C_D}, \frac{g_{m3}}{C_E}, \frac{2}{R_{LC}}, \frac{2g_{m1}}{C_F}$$

Problem 3 (30 points)

For the circuit shown below, assume that all transistors are operating in the saturation region. This circuit drives a resistive load through an ideal (infinite) AC-coupling capacitor that can be considered as a perfect short in small-signal AC analysis. Give expressions for the following and **do NOT neglect the transistor's output resistance**.

- a) Loaded small-signal gain
- b) Output resistance. Note, don't include the load resistor, R_L .



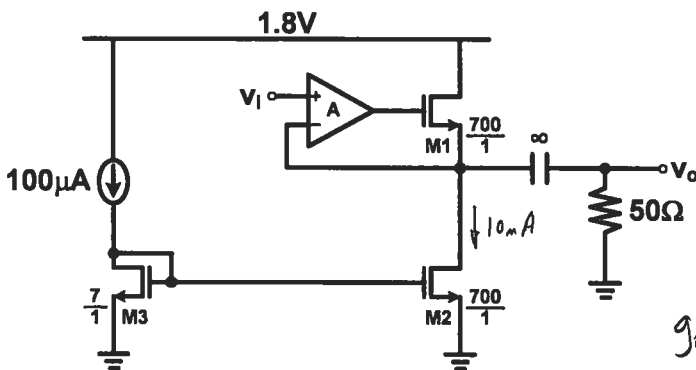
$$A_V = \frac{g_{m1} A}{g_{m1}(1+A) + g_{o1} + g_{o2} + G_L}$$

$$R_{out} = \frac{1}{g_{m1}(1+A) + g_{o1} + g_{o2}}$$

Now a more complete size circuit with the biasing details is shown that drives a 50Ω load resistor. What should the value of A be to achieve a loaded gain of $0.9V/V$?

Assume the following transistor parameters.

$$K_{PN} = \mu_n C_{ox} = 200 \mu A/V^2, V_{TH,N} = 0.4V, \lambda_N = 0.1V^{-1}$$



$$A = \frac{A_V (g_{m1} + g_{o1} + g_{o2} + G_L)}{g_{m1}(1 - A_V)}$$

$$g_{m1} = \sqrt{(200 \mu)(\frac{700}{1})(2)(10 \mu A)} = 52.9 \text{ mA/V}$$

$$g_{o1} = g_{o2} = (0.1V^{-1})(10 \mu A) = 1 \text{ mA/V}$$

$$G_L = \frac{1}{50 \Omega} = 20 \text{ mA/V}$$

$$A = \frac{0.9(52.9 \text{ mA/V} + 1 \text{ mA/V} + 1 \text{ mA/V} + 20 \text{ mA/V})}{52.9 \text{ mA/V}(1 - 0.9)} = 12.7$$

$$A \text{ for Gain} = 0.9V/V = 12.7$$

Scratch Paper