# ECEN 605 LINEAR SYSTEMS

#### Lecture 13

#### State Feedback and Observers I – Eigenvalue Assignment by State Feedback

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#### Multivariable system

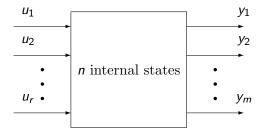


Figure 1: Input - Output System

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### Single Input Systems

Consider a single input system of form:

$$\dot{x} = Ax + bu y = C^T x.$$

If we let u(t) = 0, then

$$\dot{x} = Ax$$
  $y = C^T x.$ 

Consequently,

$$x(t) = e^{At}x(0)$$
 and  $y(t) = C^T e^{At}x(0)$ .

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Suppose that the output y(t) is unsatisfactory, then we need a controller to regulate the system. The *state feedback* problem considers the following:

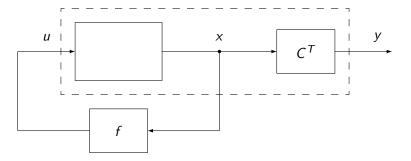


Figure 2: State Feedback Configuration

Introducing a state feedback f, we have

$$u(t) = fx(t) \implies y(t) = C^T e^{(A+bf)t} x(t)$$

so that y(t) may be satisfactory. For example,

If A is unstable, can A be stabilized by f?
 Can we find an f such that eigenvalues of A + bf (i.e., poles of the closed loop system) equal to a prescribed set of eigenvalues Λ?

This problem is the *pole assignment problem* using state feedback.

#### Problem

Given (A, b) and a desired set of eigenvalues, find f so that the eigenvalues of A + bf equal the desired set. Then the *feedback* control law

u = fx

assigns the eigenvalues of the closed loop system to the desired location.

Consider the following special case.

$$A = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & & \vdots \\ \vdots & & & \ddots & 0 \\ \vdots & & & 1 \\ a_0 & a_1 & a_2 & \cdots & a_{n-1} \end{bmatrix} \qquad b = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

and

$$\Lambda = \left[\lambda_1^d, \lambda_2^d, \cdots, \lambda_n^d\right].$$

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Let

$$f = \left[ \begin{array}{cccc} f_0 & f_1 & f_2 & \cdots & f_{n-1} \end{array} \right],$$

then

$$A+bf = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & & \vdots \\ \vdots & & & \ddots & 0 \\ \vdots & & & & 1 \\ a_0+f_0 & a_1+f_1 & a_2+f_2 & \cdots & a_{n-1}+f_{n-1} \end{bmatrix}$$

 $\mathsf{and}$ 

$$det [sl - (A + bf)] = s^{n} - (a_{n-1} + f_{n-1})s^{n-1} + \dots + (a1 + f_{1})s + (a_{0} + f_{0})$$
  
=  $(s - \lambda_{1}^{d})(s - \lambda_{2}^{d}) \cdots (s - \lambda_{n}^{d})$   
=  $s^{n} - a_{n-1}^{d}s^{n-1} + \dots + a_{1}^{d}s + a_{0}^{d}.$ 

Now we can equate the corresponding coefficients,

$$\begin{array}{rcl}
a_0 + f_0 &=& a_0^d \\
a_1 + f_1 &=& a_1^d \\
&& \vdots \\
a_{n-1} + f_{n-1} &=& a_{n-1}^d
\end{array}$$

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and solve for  $f_i$ s.

The solution consists of the following:

1) Taking an arbitrary system (A, b) and transforming it to controllable companion form by a coordinate transformation

2) Solve the *easy version* of pole assignment problem in this coordinate system

3) transform back to the original coordinates so that the same eigenvalues are obtained.

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#### When is it possible?

#### Lemma

If (A, b) is controllable, there exists a coordinate transformation T such that

$$A_{n} = T^{-1}AT = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & & \vdots \\ \vdots & & \ddots & 0 \\ \vdots & & & 1 \\ a_{0} & a_{1} & a_{2} & \cdots & a_{n-1} \end{bmatrix} \qquad b_{n} = T^{-1}b = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

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Theorem

Pole assignment by state feedback is possible iff (A, b) is a controllable pair.

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#### Proof

Suppose that (A, b) is not controllable, then we know we can separate controllable and uncontrollable parts as follows.

$$\dot{x} = Ax + bu$$

$$\Downarrow x = Tz$$

$$\dot{z} = T^{-1}ATz + T^{-1}bu$$

$$\begin{bmatrix} \dot{z}_c \\ \dot{z}_u \end{bmatrix} = \underbrace{\begin{bmatrix} A_1 & A_3 \\ 0 & A_2 \end{bmatrix}}_{\hat{A}} \begin{bmatrix} z_c \\ z_u \end{bmatrix} + \underbrace{\begin{bmatrix} b_1 \\ 0 \\ \dot{b} \end{bmatrix}}_{\hat{b}} u$$

Since

$$u = fx = fTz = \hat{f}z = \begin{bmatrix} \hat{f}_1 & \hat{f}_2 \end{bmatrix} \begin{bmatrix} z_c \\ z_u \end{bmatrix},$$
$$\begin{bmatrix} \dot{z}_c \\ \dot{z}_u \end{bmatrix} = \underbrace{\begin{bmatrix} A_1 + b_1 \hat{f}_1 & A_3 + b_1 \hat{f}_2 \\ 0 & A_2 \end{bmatrix}}_{\hat{A} + \hat{b}\hat{f}} \begin{bmatrix} z_c \\ z_u \end{bmatrix}$$

The eigenvalues of  $\hat{A}+\hat{b}\hat{f}$  are the roots of the polynomial

$$det \left[ sI - (\hat{A} + \hat{b}\hat{f}) \right] = det \left[ \begin{array}{c} sI - (A_1 + b_1\hat{f}_1) & -(A_3 + b_1\hat{f}_2) \\ 0 & sI - A_2 \end{array} \right]$$
$$= det \left[ sI - (A_1 + b_1\hat{f}_1) \right] det \left[ sI - A_2 \right].$$

As seen f has no effect on the uncontrollable part of eigenvalues that is, the eigenvalues of  $A_2$  are fixed and independent of f.

#### How to make the controllable companion transformation

The following procedure constructs a transformation matrix that coordinate transforms an arbitrary controllable system to the controllable companion form.

1)  

$$L = \begin{bmatrix} b & Ab & \cdots & A^{n-1}b \end{bmatrix}$$
2) take the last row of  $L^{-1}$  and call it  $q^T$ 
3) Construct

$$T^{-1} = \left[egin{array}{c} q^T \ q^T A \ dots \ q^T A^{n-1} \end{array}
ight]$$

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Proceeding, let a state feedback  $\hat{f}$  assign the eigenvalues of  $\hat{A} + \hat{b}\hat{f}$  to the desired locations. The last step is to find the solution

 $f = \hat{f} T^{-1}$ 

which is valid in the original coordinates, because from

$$A + bf = T(\hat{A} + \hat{b}\hat{f})T^{-1}$$

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 $\hat{A} + \hat{b}\hat{f}$  and A + bf have the same eigenvalues.

### **Multi Input Systems**

Consider a multi input system,

$$\dot{x} = Ax + Bu, \qquad A \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times m}.$$

In some cases (which cases?) we can reduce this system to a single input system by introducing a new signal

$$u = gv$$
  $g \in \mathbb{R}^m$ 

and retain controllability of the system from the new input v.

#### Multi Input Systems (cont.)

Then we have the new system which has a single input,

$$\dot{x} = Ax + Bgv := Ax + bv.$$

We now design a state feedback for this system. In general, we may have to use coordinate transformation of z = Tx.

$$\dot{z} = T^{-1}ATz + T^{-1}bv := \hat{A}z + \hat{b}v; \qquad v = \hat{f}z$$
$$= (\hat{A} + \hat{b}\hat{f})z.$$

Consequently, since

$$v = \hat{f}z = \hat{f}T^{-1}x = fx,$$

we have

$$\dot{x} = Ax + bv = (A + b \underbrace{\hat{f} T^{-1}}_{f})x = (A + B \underbrace{g \hat{f} T^{-1}}_{F})x. \quad (1)$$

## Multi Input Systems (cont.)

#### Remark

This approach of using controllable companion form can be numerically unreliable, because the controllable companion form transformation is sometimes numerically ill conditioned.

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#### **Solution Using Sylvester's Equation**

An attractive alternative method of solution is as follows.

Consider the equation

 $X^{-1}(A + BF)X = \tilde{A};$   $\tilde{A}$  has the desired set of eigenvalues.

Then,

$$\begin{array}{rcl} AX + BFX &=& X\tilde{A} \\ AX - X\tilde{A} &=& -BFX. \end{array}$$

This leads the following matrix equations:

$$AX - X\tilde{A} = -BG;$$
 given  $A$  and  $\tilde{A}$ , a choice of  $G$  (2)  
 $F = GX^{-1}.$  (3)

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### Solution Using Sylvester's Equation (cont.)

The questions that arise are:

1) Does the solution of Eq. (2) always exist? (perhaps, unique?)

- 2) Is the solution X invertible?
- 3) How to choose G?

### Solution Using Sylvester's Equation (cont.)

#### Lemma

If (A, B) is controllable, and  $(G, \tilde{A})$  is observable, then the unique solution X of eq. (2) is "almost always" nonsingular.

Based on this we develop the procedure: Procedure:

Pick à such that it has the desired eigenvalues.
 Pick G with G, Ã and solve eq. (2). If X is singular, choose a different G and repeat the process.
 If X is nonsingular, solve for F from eq. (3).