ECEN 605 LINEAR SYSTEMS

Lecture 17

State Feedback and Observers VI – Pole Placement Compensators

Pole Placement Compensators

Let the plant equation be

$$\dot{x} = Ax + Bu$$

 $y = Cx$ measured output

Let the compensator equation be

$$\dot{x}_c = A_c x_c + B_c y
u_c = C_c x_c + D_c y$$

Therefore, the plant and compensator transfer functions are

$$G_p(s) = C(sI - A)^{-1}B$$

 $G_c(s) = C_c(sI - A_c)^{-1}B_c + D_c.$

Pole Placement Compensators (cont.)

We would like to design (A_c, B_c, C_c, D_c) .

Design Problem: Choose (A_c, B_c, C_c, D_c) such that the closed loop eigenvalues assume the desired values. (General Pole Placement Compensators)

What are the closed loop eigenvalues?

$$\begin{bmatrix} \dot{x} \\ \dot{x}_c \end{bmatrix} = \underbrace{\begin{bmatrix} A + BD_cC & BC_c \\ B_cC & A_c \end{bmatrix}}_{A_{cl}} \begin{bmatrix} x \\ x_c \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} v$$

where $v + u_c = u$. If the compensator order is n_c , $n + n_c$ eigenvalues will be placed in the prescribed locations.

How do we find (A_c, B_c, C_c, D_c) ?

Single Input System

$$\dot{x} = Ax + bu
y = Cx$$

1. Let p^* be the smallest integer j such that

$$\operatorname{Rank} \left[\begin{array}{c} C \\ CA \\ \vdots \\ CA^{j} \end{array} \right] = n(\operatorname{full\ rank})$$

 p^* exists and is unique if (C, A) is observable, which we assume. Add p $(p \ge p^*)$ integrators to the system as follows:

$$\begin{array}{rcl} u & = & u_1 \\ \dot{u}_1 & = & u_2 \\ \dot{u}_2 & = & u_3 \\ \vdots & & & \\ \dot{u}_p & = & v \end{array}$$

Introduce the extended state vector

$$ar{x} := \left[egin{array}{c} x \\ u_1 \\ u_2 \\ \vdots \\ u_p \end{array} \right]$$

and consider the augmented system

$$\dot{\bar{x}} = \bar{A}\bar{x} + \bar{b}v$$

where

$$ar{A} = \left[egin{array}{cccccc} A & b & 0 & \cdots & \cdots & 0 \\ 0 & 0 & 1 & & & dots \\ dots & & 0 & 1 & & dots \\ dots & & & 0 & 1 & & dots \\ dots & & & & \ddots & 0 \\ dots & & & & & 1 \\ 0 & \cdots & \cdots & \cdots & \cdots & 0 \end{array}
ight] \qquad ar{b} = \left[egin{array}{c} 0 \\ dots \\ dots \\ 0 \\ 1 \end{array}
ight]$$

We note that (\bar{A}, \bar{b}) is controllable if (A, b) is so.

2. Find the state feedback control

$$v = \bar{f}\bar{x}$$

so that $\bar{A} + \bar{b}\bar{f}$ has a prescribed set Λ of n+p eigenvalues in the LHP. Write

$$\underbrace{\overline{f}}_{1\times n+p} = \left[\begin{array}{ccc} \underbrace{f_0}_{1\times n} & f_1 & f_2 & \cdots & f_p \end{array} \right]$$

3. Introduce the feedback compensator with transfer function (input y; output u).

$$C(s) = \frac{\beta_0 + \beta_1 s + \dots + \beta_p s^p}{s^p - \alpha_{p-1} s^{p-1} - \dots - \alpha_1 s - \alpha_0}$$

where C(s) is $1 \times m$, β_i are $1 \times m$, and α_i are scalar. A state space realization of this compensator is

$$A_{c} = \begin{bmatrix} 0 & & & \alpha_{0} \\ 1 & & & \alpha_{1} \\ 0 & \ddots & & \vdots \\ \vdots & & \ddots & \vdots \\ 0 & & 1 & \alpha_{p-1} \end{bmatrix} \qquad B_{c} = \begin{bmatrix} \beta_{0} + \alpha_{0}\beta_{p} \\ \beta_{1} + \alpha_{1}\beta_{p} \\ \vdots \\ \vdots \\ \beta_{p-1} + \alpha_{p-1}\beta_{p} \end{bmatrix}$$

$$c_c = \begin{bmatrix} 0 & 0 & \cdots & 0 & 1 \end{bmatrix}$$
 $D_c = \beta_p$.

4. Calculate the β_i from the equation

$$\begin{bmatrix} \beta_0 & \beta_1 & \cdots & \beta_p \end{bmatrix} \begin{bmatrix} C \\ CA \\ \vdots \\ CA^p \end{bmatrix} = f_0.$$

This equation has a solution because of the observability assumption.

5. Calculate the α_i from the equations

$$\beta_{p}CA^{p-1}b + \beta_{p-1}CA^{p-2}b + \dots + \beta_{1}Cb + \alpha_{0} = f_{1}$$
 $\beta_{p}CA^{p-2}b + \beta_{p-1}CA^{p-3}b + \dots + \beta_{2}Cb + \alpha_{1} = f_{2}$
 \vdots

$$\beta_{p}Cb + \alpha_{p-1} = f_{p}$$

This determines all the compensator parameters. It remains to be proved that the closed loop poles (or eigenvalues) are exactly equal to the set Λ of n + p eigenvalues that were assigned in Step 2.

Theorem

The compensator designed above assigns the closed loop eigenvalues to the set Λ .

Proof

We will show that the closed loop system matrix A_{cl} is similar to $\bar{A} + \bar{b}\bar{f}$. Let us write out

$$A_{cl} = \begin{bmatrix} A + bD_{c}C & bc_{c} \\ B_{c}C & A_{c} \end{bmatrix}$$

$$= \begin{bmatrix} A + b\beta_{p}C & \vdots & 0 & 0 & \cdots & 0 & b \\ (\beta_{0} + \alpha_{0}\beta_{p})C & \vdots & 0 & 0 & \cdots & 0 & \alpha_{0} \\ (\beta_{1} + \alpha_{1}\beta_{p})C & \vdots & 1 & & & \alpha_{1} \\ \vdots & \vdots & \ddots & & \vdots \\ \vdots & \vdots & \ddots & & \vdots \\ (\beta_{p-1} + \alpha_{p-1}\beta_{p})C & \vdots & 0 & \cdots & 0 & 1 & \alpha_{p-1} \end{bmatrix}$$

Single Output System

1. Let q^* be the least integer j such that

$$\operatorname{Rank} \left[B \ AB \ A^2B \ \cdots \ A^jB \right] = n$$

 q^* exists and is unique when (A, B) is controllable. The compensator order can be any $q \ge q^*$.

2. Form the pair (\bar{c}, \bar{A}) , $\bar{A} \in \mathbb{R}^{(n+q)\times(n+q)}$, $\bar{c} \in \mathbb{R}^{1\times(n+q)}$ with

$$\bar{A} = \begin{bmatrix} A & 0 & \cdots & \cdots & \cdots & 0 \\ c & 0 & & & & & 0 \\ 0 & 1 & & & & \vdots \\ 0 & 0 & 1 & & & \vdots \\ \vdots & & & \ddots & & \vdots \\ \vdots & & & \ddots & & \vdots \\ 0 & 0 & \cdots & \cdots & 0 & 1 & 0 \end{bmatrix}$$

$$\bar{c} = \begin{bmatrix} \underline{0} & 0 & \cdots & \cdots & 0 & 1 \end{bmatrix}$$

 (\bar{c}, \bar{A}) is observable when (c, A) is.

3. Using pole placement, find

$$\bar{L} = \left[\begin{array}{c} L_0 \\ l_1 \\ \vdots \\ l_q \end{array} \right]$$

so that $\bar{A}+\bar{L}\bar{c}$ has the desired set Λ of n+q eigenvalues in the left half of the complex plane.

4. Consider the feedback compensator with transfer function

$$\frac{\gamma_0 + \gamma_1 s + \dots + \gamma_q s^q}{s^q - \alpha_{q-1} s^{q-1} - \dots - \alpha_1 s - \alpha_0} \qquad \gamma_i \in \mathbb{R}^v, \quad \alpha_i \text{ is scalar}$$

A state space realization of this is

$$A_c = \left[egin{array}{ccccc} 0 & 1 & 0 & \cdots & 0 \ dots & 1 & & dots \ dots & & \ddots & dots \ dots & & & 1 \ dots & & & 1 \ lpha_0 & lpha_1 & \cdots & \cdots & lpha_{q-1} \end{array}
ight] \qquad b_c = \left[egin{array}{c} 0 \ 0 \ dots \ 0 \ dots \ 1 \end{array}
ight]$$

$$C_c = [\gamma_0 + \alpha_0 \gamma_q \quad \gamma_1 + \alpha_1 \gamma_q \quad \cdots \quad \gamma_{q-1} + \alpha_{q-1} \gamma_q] \qquad D_c = \gamma_q$$

5. Calculate γ_i , $i = 0, 1, \dots, q$ from

$$\begin{bmatrix} B & AB & \cdots & A^qB \end{bmatrix} \begin{bmatrix} \gamma_0 \\ \gamma_1 \\ \vdots \\ \gamma_q \end{bmatrix} = L_0$$

This equation has a solution because the matrix on the left has full column rank.

6. Calculate the α_i , $i = 0, 1, \dots, q-1$ from

$$cB\gamma_{1} + cAB\gamma_{2} + \dots + cA^{q-1}B\gamma_{q} + \alpha_{0} = I_{1}$$

$$cB\gamma_{2} + \dots + cA^{q-2}B\gamma_{q} + \alpha_{1} = I_{2}$$

$$\vdots$$

$$cB\gamma_{q} + \alpha_{q-1} = I_{q}$$

This determines all the compensator parameters.

Theorem

The compensator designed by the above procedure assigns the closed loop eigenvalues to Λ .

Proof

By construction we have that the eigenvalues of $\bar{A} + \bar{L}\bar{c}$ equal to Λ . We will show that $\bar{A} + \bar{L}\bar{c}$ is similar to the closed loop matrix A_{cl} . For this write out A_{cl} :

$$\bar{A}_{cl} = \begin{bmatrix} A + B\gamma_q c & B(\gamma_0 + \alpha_0 \gamma_q) & B(\gamma_1 + \alpha_1 \gamma_q) & \cdots & \cdots & B(\gamma_{q-1} + \alpha_{q-1} \gamma_q) \\ 0 & 0 & 1 & & \vdots \\ \vdots & & & \ddots & 0 \\ \vdots & & & & \ddots & 0 \\ \vdots & & & & & \ddots & 0 \\ \vdots & & & & & \ddots & \ddots & \vdots \\ \vdots & & & & \ddots & \ddots & \vdots \\ 0 & 1 & & & \vdots & I_2 \\ \vdots & & & & \ddots & \ddots & \vdots \\ \vdots & & & & \ddots & 0 & \vdots \\ 0 & \cdots & \cdots & 0 & 1 & I_q \end{bmatrix}$$

$$ar{A} + ar{L}ar{c} = \left[egin{array}{cccccc} A & 0 & 0 & \cdots & 0 & L_0 \\ c & 0 & 0 & & 0 & l_1 \\ 0 & 1 & & dots & l_2 \\ dots & & \ddots & & dots & dots \\ dots & & \ddots & 0 & dots \\ 0 & \cdots & \cdots & 0 & 1 & l_g \end{array}
ight.$$

Now it can be verified that

$$QA_{cl} = (\bar{A} + \bar{L}\bar{c})Q$$

where

$$Q = \begin{bmatrix} I_n & P_1 & P_2 & \cdots & \cdots & P_q \\ 0 & -I_2 & -I_3 & \cdots & -I_q & 1 \\ 0 & -I_3 & & -I_q & 0 & 0 \\ \vdots & & & & \vdots \\ \vdots & -I_q & & & \vdots \\ 0 & 1 & 0 & \cdots & \cdots & 0 \end{bmatrix}$$

$$P_i = -B\gamma_i - AB\gamma_{i+1} - \cdots - A^{q-i}B\gamma_q, \qquad i = 1, 2, \cdots, q.$$

$$P_i = -B\gamma_i - AB\gamma_{i+1} - \cdots - A^{q-i}B\gamma_q, \qquad i = 1, 2, \cdots, q.$$

Example

$$A = \begin{bmatrix} 0 & 1 & 1 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 1 \end{bmatrix} \qquad b = \begin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \end{bmatrix} \qquad c = \begin{bmatrix} 0 & 0 & 0 & 1 \end{bmatrix}$$

Since n = 4, we need to add n - 1 = 3 integrators.

$$u_{1} = u, \quad \dot{u}_{1} = u_{2}, \quad \dot{u}_{2} = u_{3}, \quad \dot{u}_{3} = v$$

$$\begin{bmatrix} \dot{x} \\ \dot{u}_{1} \\ \dot{u}_{2} \\ \dot{u}_{3} \end{bmatrix} = \underbrace{\begin{bmatrix} A & b & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}}_{\bar{A}} \begin{bmatrix} x \\ u_{1} \\ u_{2} \\ u_{3} \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix}}_{\bar{b}} v$$

$$v = f_0 x + f_1 u_1 + f_2 u_2 + f_3 u_3 = \underbrace{\begin{bmatrix} f_0 & f_1 & f_2 & f_3 \end{bmatrix}}_{f} \begin{bmatrix} x & u_1 \\ u_2 \\ u_3 \end{bmatrix}$$

and

$$f_0 = [f_{01} \quad f_{02} \quad f_{03} \quad f_{04}].$$

Now suppose we assign the closed eigenvalues at -1, -2, -3, -4, -5, -6, and -7. Then

$$\hat{\bar{A}} + \hat{\bar{b}}\hat{f} = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ \hat{f}_{01} & \hat{f}_{02} & \hat{f}_{03} & 1 + \hat{f}_{04} & -2 - \hat{f}_{1} & \hat{f}_{2} & 2 + \hat{f}_{3} \end{bmatrix}.$$

Once we find \hat{f} , we have

$$f = \hat{f} T^{-1} = \begin{bmatrix} f_0 & f_1 & f_2 & f_3 \end{bmatrix}.$$

Now let the controller transfer function be

$$C(s) = \frac{\beta_0 + \beta_1 s + \beta_2 s^2 + \beta_3 s^3}{s^3 - \alpha_2 s^2 - \alpha_1 s - \alpha_0}$$

then solve the following matrix equation to get the numerator coefficits.

$$\begin{bmatrix} \beta_0 & \beta_1 & \beta_2 & \beta_3 \end{bmatrix} \begin{bmatrix} c \\ cA \\ cA^2 \\ cA^3 \end{bmatrix} = \begin{bmatrix} f_{01} & f_{02} & f_{03} & f_{04} \end{bmatrix} = f_0.$$

The denominator coefficients are obtained from the following.

$$\alpha_0 + \beta_1 cb + \beta_2 cAb + \beta_3 cA^2b = f_1$$

$$\alpha_1 + \beta_2 cb + \beta_2 cAb = f_2$$

$$\alpha_2 + \beta_3 cb = f_3.$$

Finally, we have

$$C(s) = \frac{-15768 - 7466s - 12398s^2 - 12154s^3}{s^3 + 30s^2 + 382s + 2722}.$$

Example

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 1 & 0 & 1 \\ 0 & 1 & 0 \end{bmatrix} \qquad B = \begin{bmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix} \qquad c = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$

Find a pole placement compensator so that all closed loop poles are assigned at -2.

Since

$$Rank [B AB] = 3,$$

the order of compensator is 1. Thus, we form the argument system

$$\bar{A} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 0 & 0 & 0 \end{bmatrix} \qquad \bar{c} = \begin{bmatrix} 0 & 0 & 0 & 1 \end{bmatrix}.$$

Since (c, A) is observable, (\bar{c}, \bar{A}) is onservable. Now let us find L so that $\lambda(\bar{A} + L\bar{c}) = \Lambda$.

$$\bar{A} + L\bar{c} = \begin{bmatrix}
0 & 1 & 0 & 0 \\
1 & 0 & 1 & 0 \\
0 & 1 & 0 & 0 \\
1 & 0 & 0 & 0
\end{bmatrix} + \begin{bmatrix}
l_{01} \\
l_{02} \\
l_{03} \\
l_{1}
\end{bmatrix} \begin{bmatrix}
0 & 0 & 0 & 1
\end{bmatrix}$$

$$= \begin{bmatrix}
0 & 1 & 0 & l_{01} \\
1 & 0 & 1 & l_{02} \\
0 & 1 & 0 & l_{03} \\
1 & 0 & 0 & l_{1}
\end{bmatrix}$$

We found

$$L = \begin{bmatrix} -26 \\ -48 \\ -42 \\ -8 \end{bmatrix} L_0$$

$$\begin{bmatrix} L_0 \\ -42 \\ -8 \end{bmatrix} L_1$$

Now let the compensator transfer function be

$$C(s) = \frac{\gamma_0 + \gamma_1 s}{s - \alpha_0}.$$

We solve the following matrix equation

$$\left[\begin{array}{ccc} B & AB \end{array} \right] \left[\begin{array}{c} \gamma_0 \\ \gamma_1 \end{array} \right] = \left[\begin{array}{ccc} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 0 \end{array} \right] \left[\begin{array}{c} \gamma_{01} \\ \gamma_{02} \\ \gamma_{11} \\ \gamma_{12} \end{array} \right] = \left[\begin{array}{c} -26 \\ -48 \\ -42 \end{array} \right] = L_0.$$

Thus we have

$$\gamma_0 = \left[egin{array}{c} -26 \\ -42 \end{array}
ight] \qquad \gamma_1 = \left[egin{array}{c} -1 \\ -47 \end{array}
ight].$$

Now we compute α_0

$$cB\gamma_1 + \alpha_0 = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} -1 \\ -47 \end{bmatrix} + \alpha_0 = L_1 = -8.$$

Thus, we have

$$\alpha_0 = -7.$$

A state space representation of the compensator is

$$A_c = -7$$
, $b_c = 1$, $C_c = \gamma_0 + \alpha_0 \gamma_1 = \begin{bmatrix} -19 \\ 287 \end{bmatrix}$, $D_c = \begin{bmatrix} -1 \\ -47 \end{bmatrix}$.

The composite system is

$$\begin{bmatrix} \dot{x} \\ \dot{x}_c \end{bmatrix} = \begin{bmatrix} A + BD_cC & BC_c \\ B_cC & A_c \end{bmatrix} \begin{bmatrix} x \\ x_c \end{bmatrix}$$
$$= \begin{bmatrix} -1 & 1 & 0 & -19 \\ 1 & 0 & 1 & 0 \\ -47 & 1 & 0 & 287 \\ 1 & 0 & 0 & -7 \end{bmatrix} \begin{bmatrix} x \\ x_c \end{bmatrix}.$$